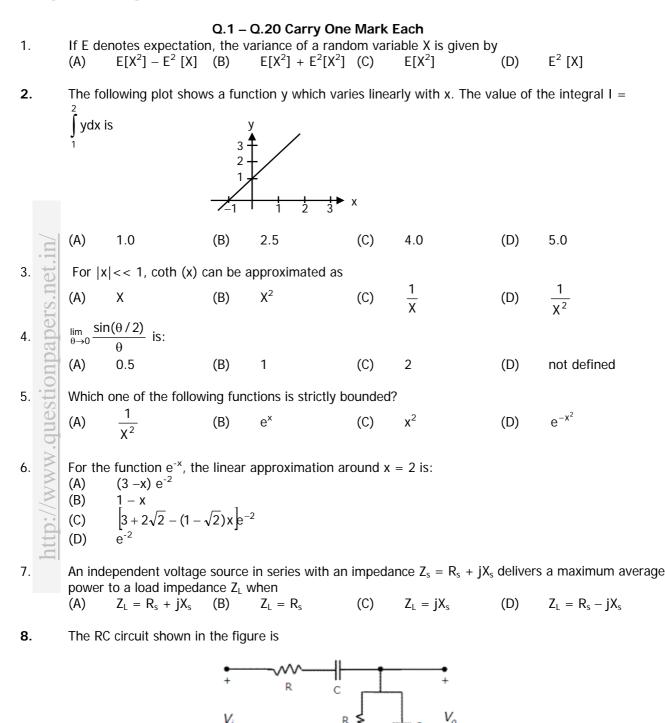
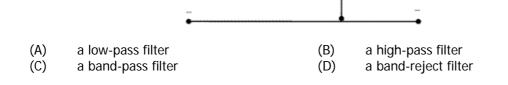
# GATE question papers: Electronics and Communication Engineering 2007 (EC)

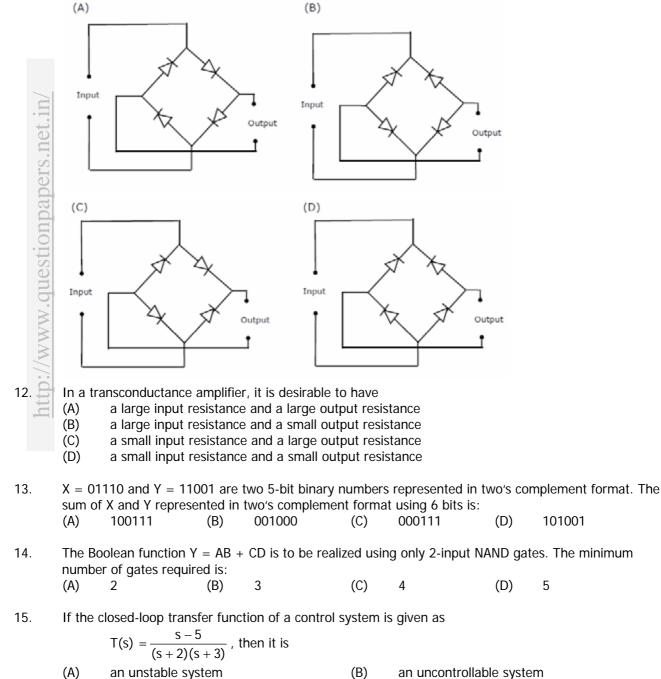




9. The electron and hole concentrations in an intrinsic semiconductor are n<sub>i</sub> per cm<sup>3</sup> at 300 K. Now, if acceptor impurities are introduced with a concentration of N<sub>A</sub> per cm<sup>3</sup> (where N >> n), the electron concentration per cm<sup>3</sup> at 300 K will be

(A) 
$$n_i$$
 (B)  $n_i + N_A$  (C)  $N_A - n_i$  (D)  $\frac{n_i^-}{N_A}$ 

- 10. In a p<sup>+</sup> n junction diode under reverse bias, the magnitude of electric field is maximum at
  - (A) the edge of the depletion region on the p-side
  - (B) the edge of the depletion region on the n-side
  - (C) the  $p^+$  n junction
  - (D) the centre of the depletion region on the n-side
- 11. The correct full wave rectifier circuit is:



(C) a minimum phase system (D) a non-minimum phase system

16. If the Laplace transform of a signal y (t) is 
$$Y(s) = \frac{1}{s(s-1)}$$
, then its final value is:  
(A) -1 (B) 0 (C) 1 (D) unbounded

- 17. If R ( $\tau$ ) is the autocorrelation function of a real, wide-sense stationary random process, then which of the following is NOT true?
  - $R(\tau) = R(-\tau)$ (A)
  - (B)  $|\mathsf{R}(\tau)| \leq \mathsf{R}(0)$
  - (C)  $R(\tau) = -R(-\tau)$
  - (D) The mean square value of the process is R (0)
- 18. If S (f) is the power spectral density of a real, wide-sense stationary random process, then which of the following is ALWAYS true?

(A) 
$$S(0) \ge S(f)$$
 (B)  $S(f) \ge 0$   
(C)  $S(-f) = -S(f)$  (D)  $\int_{-\infty}^{\infty} S(f) df = 0$ 

A plane wave of wavelength  $\lambda$  is traveling in a direction making an angle 30° with positive x-axis and 90° with positive y-axis. The  $\vec{E}$  field of the plane wave can be represented as (E<sub>0</sub> is constant)

(A) 
$$\vec{E} = \hat{y}E_0e^{j\left(\omega t - \frac{\sqrt{3}\pi}{\lambda}x - \frac{\pi}{\lambda}z\right)}$$
  
(B)  $\vec{E} = \hat{y}E_0e^{j\left(\omega t - \frac{\pi}{\lambda}x - \frac{\sqrt{3}\pi}{\lambda}z\right)}$   
(C)  $\vec{E} = \hat{y}E_0e^{j\left(\omega t + \frac{\sqrt{3}\pi}{\lambda}x + \frac{\pi}{\lambda}z\right)}$   
(D)  $\vec{E} = \hat{y}E_0e^{j\left(\omega t - \frac{\pi}{\lambda}x + \frac{\sqrt{3}\pi}{\lambda}z\right)}$ 

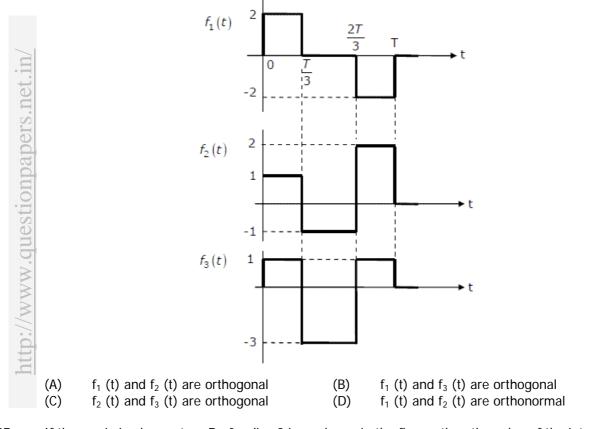
nttp://www.guestionpapers.net If C is a closed curve enclosing a surface S, then the magnetic field intensity  $\vec{H}$ , the current density 20.  $\overrightarrow{J}$  and the electric flux density  $\overrightarrow{D}$  are related by

(A) 
$$\iint_{S} \vec{H}.d\vec{s} = \oint_{C} \left( \vec{J} + \frac{\partial \vec{D}}{\partial t} \right).d\vec{l}$$
(B) 
$$\int_{C} \vec{H}.d\vec{l} = \oiint_{S} \left( \vec{J} + \frac{\partial \vec{D}}{\partial t} d \right).d\vec{s}$$
(C) 
$$\oiint_{S} \vec{H}.d\vec{s} = \int_{C} \left( \vec{J} + \frac{\partial \vec{D}}{\partial t} \right).d\vec{l}$$
(D) 
$$\oint_{C} \vec{H}.d\vec{l} = \iint_{S} \left( \vec{J} + \frac{\partial \vec{D}}{\partial t} \right).d\vec{s}$$

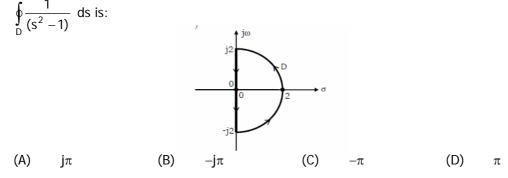
#### Q. 21 - Q.75 carry Two Marks Each

- It is given that X1, X2.....XM are M non-zero, orthogonal vectors. The dimension of the vector space 21. spanned by the 2M vectors  $X_1$ ,  $X_2$ .....XM,  $-X_1$ ,  $-X_2$ ....-XM is: (A) 2M (B) M + 1
  - (C) Μ (D) dependent on the choice of  $X_{11}$ ,  $X_{2}$ .....XM
- Consider the function f (x) =  $x^2 x 2$ . The maximum value of f (x) in the closed interval [-4, 4] is: 22. -2.25(A) 10 (B) 10 (C) (D) indeterminate
- An examination consists of two papers, Paper 1 and Paper 2. The probability of failing in Paper 1 is 0.3 23. and that in Paper 2 is 0.2. Given that a student has failed in Paper 2, the probability of failing in Paper 1 is 0.6. The probability of a student failing in both the papers is:
  - (D) (A) 0.5 (B) 0.18 0.12 0.06 (C)

- 24. The solution of the differential equation  $k^2 \frac{d^2y}{dx^2} = y y_2$  under the boundary conditions (i)  $y = y_1$  at
  - x = 0 and (ii)  $y = y_2$  at  $x = \infty$ , where k,  $y_1$  and  $y_2$  are constants, is(A)  $y = (y_1 y_2) \exp(-x/k^2) + y_2$ (B)  $y = (y_2 y_1) \exp(-x/k) + y_1$ (C)  $y = (y_1 y_2) \sinh(x/k) + y_1$ (D)  $y = (y_1 y_2) \exp(-x/k) + y_2$
- 25. The equation  $x^3 x^2 + 4x 4 = 0$  is to be solved using the Newton-Raphson method. If x = 2 is taken as the initial approximation of the solution, then the next approximation using this method will be:
  - (A)  $\frac{2}{3}$  (B)  $\frac{4}{3}$  (C) 1 (D)  $\frac{3}{2}$
- 26. Three functions  $f_1$  (t),  $f_2$  (t) and  $f_3$  (t), which are zero outside the interval [0, T], are shown in the figure. Which of the following statements is correct?

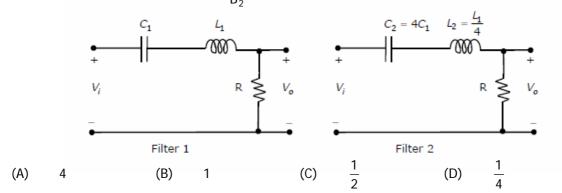


27. If the semi-circular contour D of radius 2 is as shown in the figure, then the value of the integral  $\oint \frac{1}{\sqrt{1-1}} ds$  is:

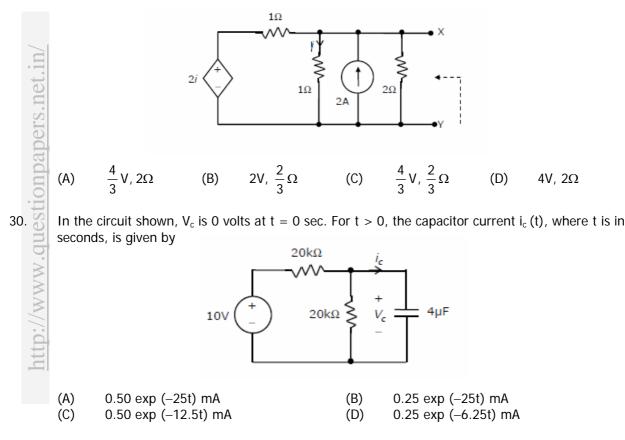


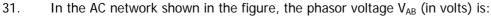
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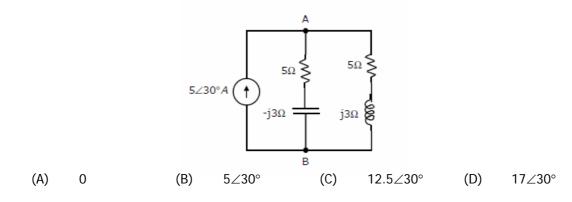
28. Two series resonant filters are as shown in the figure. Let the 3-dB bandwidth of Filter 1be B<sub>1</sub> and that of Filter 2 be B<sub>2</sub>. The value of  $\frac{B_1}{B_2}$  is:



29. For the circuit shown in the figure, the Thevenin voltage and resistance looking into X-Y are:





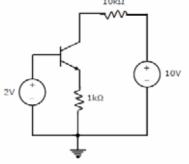


32. A p<sup>+</sup> n junction has a built-in potential of 0.8 V. The depletion layer width at a reverse bias of 1.2V is 2 m. For a reverse bias of 7.2 V, the depletion layer width will be:
(A) 4 m
(B) 4.9 m
(C) 8 m
(D) 12 m

33. Group I lists four types of p-n junction diodes. Match each device in Group I with one of the option in Group II to indicate the bias condition of that device in its normal mode of operation.

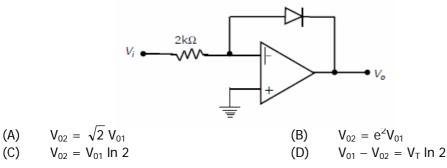
		Group	II to Ind		olas cor Group		r that dev	lice in i	Group I		r operat	ion.	
		(A) (B) (C) (D)	P - 1 P - 2 P - 2 P - 2 P - 2	Q - 2 Q - 1 Q - 2	r cell ER diod		de	• •	ward bias verse bias				
	34.		C current ort factor	0	of a BJ	T is 50. <i>A</i>	Assuming	that th	e emitter	injectior	n efficiei	ncy is 0.99	5, the base
	35. Jet. 10/	(A) Group	0.980	ur differer	(B) nt semio	0.985 conducto	or devices	(C) 5. Match	0.990 n each dev	vice in G	(D) roup I v	0.995 vith its cha	racteristic
	rs.n	proper		Group	İ			Group	П				
36.	tionpape			(P) BJT (Q) MOS (R) LASI (S) JFET	ER diod		(1) Pop (2) Pinc (3) Earl (4) Flat	h-off vo y effect	Ū				
	ww.questionpapers.net.	(A) (B) (C) (D)	P - 3 P - 1 P - 3 P - 3	Q - 4 Q - 4	R - 4 R - 3 R - 1 R - 1	S - 2 S - 2 S - 2 S - 4							
	http://wv	For the		p circuit s	hown ii ۱۸۶ ۱۸۶ ۱۸۶		ure, $V_0$ is $2k\Omega$ $-\sqrt{-}$ + + $k\Omega$		• V <sub>o</sub>				
		(A)	-2 V		(B)	-1 V		(C)	–0.5 V		(D)	0.5 V	
	37.	transis	tor is vei	cuit shown ry large a e BJT is:							10kΩ		
		(A)	cut-off						Г	$\neg$		(† )	

- (A) cut-off
- (B) saturation
- (C) normal active
- (D) reverse active

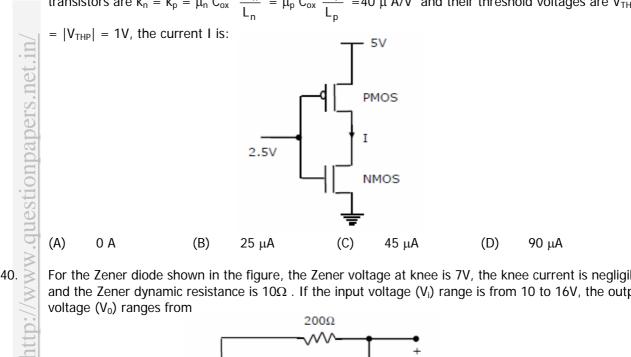


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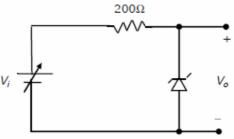
38. In the Op-Amp circuit shown, assume that the diode current follows the equation  $I = I_s \exp (V/V_T)$ . For  $V_i = 2V$ ,  $V_0 = V_{01}$ , and for  $V_i = 4V$ ,  $V_0 = V_{02}$ . The relationship between  $V_{01}$  and  $V_{02}$  is:



39. In the CMOS inverter circuit shown, if the transconductance parameters of the NMOS and PMOS transistors are  $k_n = k_p = \mu_n C_{ox}$   $\frac{W_n}{L_n} = \mu_p C_{ox} \frac{W_p}{L_n} = 40 \ \mu \text{ A/V}^2$  and their threshold voltages are  $V_{THM}$ 



For the Zener diode shown in the figure, the Zener voltage at knee is 7V, the knee current is negligible and the Zener dynamic resistance is  $10\Omega$ . If the input voltage (V<sub>i</sub>) range is from 10 to 16V, the output voltage (V<sub>0</sub>) ranges from



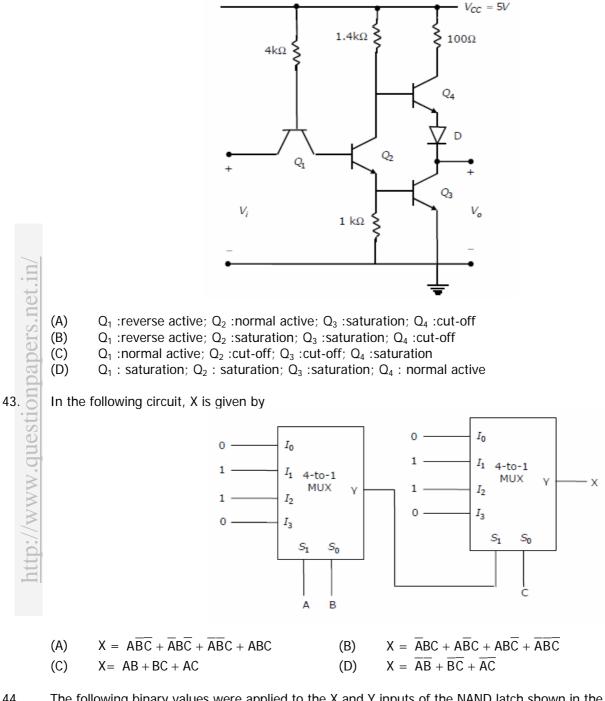
(A)	7.00 to 7.29 V	(B)	7.14 to 7.29 V
(C)	7.14 to 7.43 V	(D)	7.29 to 7.43 V

40.

The Boolean expression  $y = \overline{ABCD} + \overline{ABCD} + \overline{ABCD} + A\overline{BCD}$  can be minimized to 41.  $y = \overline{A}\overline{B}\overline{C}D + \overline{A}\overline{B}\overline{C} + \overline{A}\overline{C}D$ (A) (B)  $y = \overline{ABCD} + BC\overline{D} + A\overline{B}\overline{C}\overline{D} + AB\overline{C}D$ 

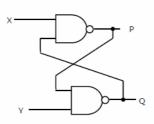
 $y = \overline{A}BC\overline{D} + \overline{B}\overline{C}D + A\overline{B}\overline{C}D$  $y = \overline{A}BC\overline{D} + \overline{B}\overline{C}D + AB\overline{C}\overline{D}$ (C) (D)

42. The circuit diagram of a standard TTL NOT gate is shown in the figure. When  $V_i = 2.5 V$ , the modes of operation of the transistors will be:



44. The following binary values were applied to the X and Y inputs of the NAND latch shown in the figure in the sequence indicated below:

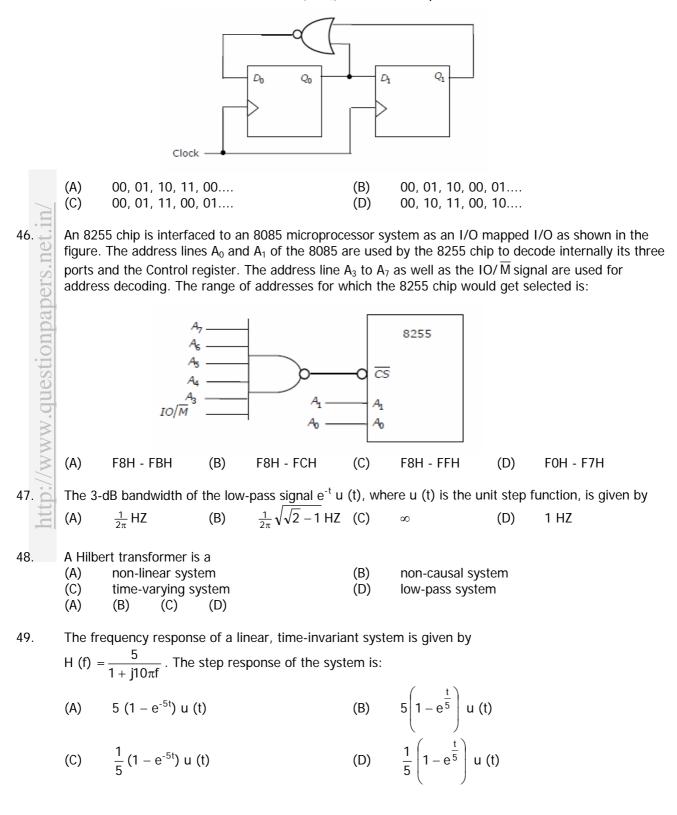
$$X = 0, Y = 1;$$
  $X = 0, Y = 0;$   $X = 1, Y = 1.$ 



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(A)	P = 1, Q = 0;	P = 1, Q = 0;	P = 1, Q = 0 c	or $P = 0, Q = 1$
(B)	P = 1, Q = 0;	P = 0, Q = 1; or	P = 0, Q = 1;	P = 0, Q = 1
(C)	P = 1, Q = 0;	P = 1, Q = 1;	P = 1, Q = 0 (	or $P = 0, Q = 1$
(D)	P = 1, Q = 0;	P = 1, Q = 1;		P = 1, Q = 1

45. For the circuit shown, the counter state  $(Q_1Q_0)$  follows the sequence



50. A 5-point sequence [n] is given as x [-3] = 1, x[-2] = 1, x[-1] = 0, x[-0] = 5, x[1] = 1, Let X ( $e^{j\omega}$ ) denote the discrete-time Fourier transform of x [n]. The value of  $\omega \int X(e^{j\omega})d\omega$  is: (A) 5 (B) (C) (D) 10π 16π  $5 + j10\pi$ The z-transform X [z] of a sequence x [n] is given by X [z] =  $\frac{0.5}{1-2z^{-1}}$ . It is given that the region of 51. convergence of X [z] includes the unit circle. The value of x [0] is: (A) -0.5 (B) 0 (C) 0.25 (D) 0.5 52. A control system with a PD controller is shown in the figure. If the velocity error constant  $K_V = 1000$ and the damping ratio  $\zeta = 0.5$ , then the values of K<sub>P</sub> and K<sub>D</sub> are: 100  $K_P + K_D s$ s(s+10)tionpapers.net.in  $\begin{array}{l} K_{P} \,=\,\, 100, \, K_{D} \,=\! 0.9 \\ K_{P} \,=\,\, 10, \, K_{D} \,=\! 0.9 \end{array}$ (B) (D) (A)  $K_P = 100, K_D = 0.09$ (C)  $K_P = 10, K_D = 0.09$ 

The transfer function of a plant is T (s) =  $\frac{5}{(s+5)(s^2+s+1)}$ . The second-order approximation of T (s) 53. June Www.//:ott using dominant pole concept is:

(A)	$\frac{1}{(s+5)(s+1)}$	(B)	$\frac{5}{(s+5)(s+1)}$
(C)	$\frac{5}{s^2 + s + 1}$	(D)	$\frac{1}{s^2 + s + 1}$

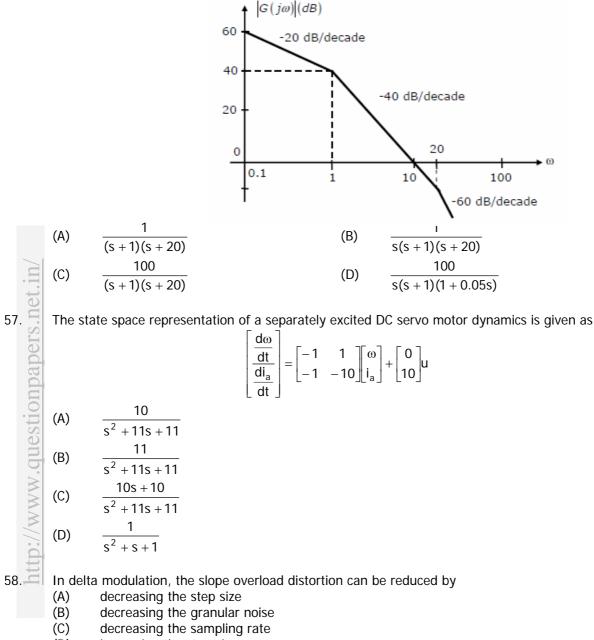
54.

The open-loop transfer function of a plant is given as G (s) =  $\frac{1}{s^2 - 1}$ . If the plant is operated in a unity feedback configuration, then the lead compensator that an stabilize this control system is:

(A)	$\frac{10(s-1)}{s+2}$	(B)	$\frac{10(s+4)}{s+2}$
(C)	$\frac{10(s+1)}{s+10}$	(D)	$\frac{2(s+2)}{s+10}$

A unity feedback control system has an open-loop transfer function G (s) =  $\frac{K}{s(s^2 + 7s + 12)}$ . The gain 55. K for which s = -1 + j1 will lie on the root locus of this system is: (A) (B) 5.5 (C) 6.5 (D) 10

56. The asymptotic Bode plot of a transfer function is as shown in the figure. The transfer function G (s) corresponding to this Bode plot is:

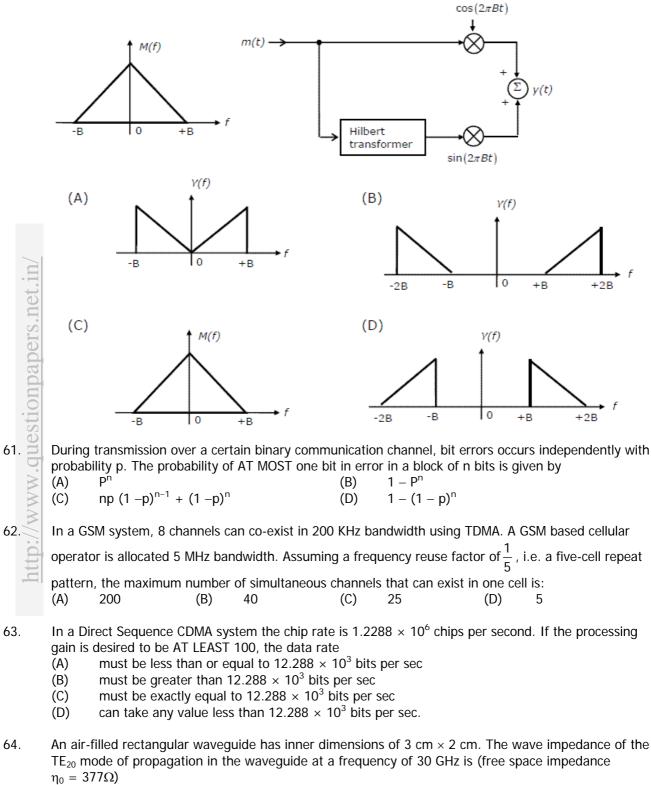


(D) increasing the step size

59. The raised cosine pulse p (t) is used for zero ISI in digital communications. The expression for p (t) with unity roll-off factor is given by p (t) =  $\frac{\sin 4\pi W t}{4\pi W t (1 - 16W^2 t^2)}$ .

The value of p (t) at t  $\frac{1}{4W}$  is: (A) -0.5 (B) 0 (C) 0.5 (D)  $\infty$ 

60. In the following scheme, if the spectrum M (f) of m (t) is as shown, then the spectrum Y (f) of y (t) will be:



(A)  $308 \Omega$  (B)  $355 \Omega$  (C)  $400 \Omega$  (D)  $461 \Omega$ 

65. The H field (in A/m) of a plane wave propagating in free space is given by

$$\vec{H} = \hat{X} \frac{5\sqrt{3}}{\eta_0} \cos(\omega t - \beta z) + \hat{y} \frac{5}{\eta_0} \sin\left(\omega t - \beta z + \frac{\pi}{2}\right).$$

The time average power flow density in watts is:

(A) 
$$\frac{\eta_0}{100}$$
 (B)  $\frac{100}{\eta_0}$  (C)  $50\eta_0^2$  (D)  $\frac{50}{\eta_0}$ 

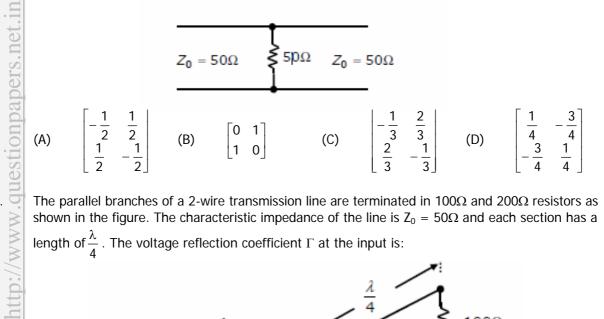
The E field in rectangular waveguide of inner dimensions  $a \times b$  is given by 66.

$$\vec{E} = \frac{\omega \mu}{h^2} \left( \frac{\pi}{a} \right) H_0 \, \sin \left( \frac{2\pi x}{a} \right)^2 \, \sin(\omega t - \beta z) \hat{y} \; . \label{eq:eq:expansion}$$

Where H<sub>0</sub> is a constant, and a and b are the dimensions along the x-axis and the y-axis respectively. The mode of propagation in the waveguide is:

(D) (A) TE<sub>20</sub> (C) TM<sub>20</sub> **TE**<sub>10</sub> (B)  $TM_{11}$ 

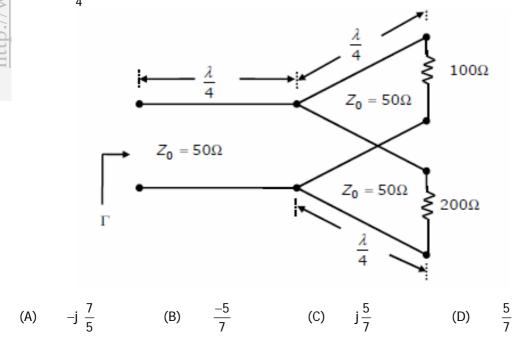
67. A load of 50 $\Omega$  is connected in shunt in a 2-wire transmission line of  $Z_0 = 50\Omega$  as shown in the figure. The 2-port scattering parameter matrix (S-matrix) of the shunt element is:



The parallel branches of a 2-wire transmission line are terminated in  $100\Omega$  and  $200\Omega$  resistors as shown in the figure. The characteristic impedance of the line is  $Z_0 = 50\Omega$  and each section has a

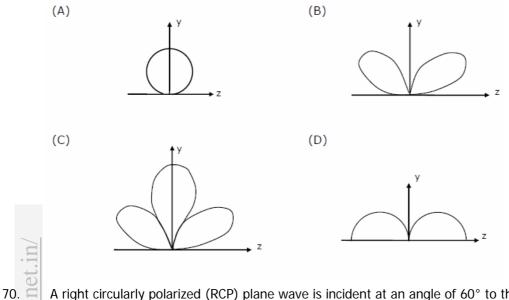
length of  $\frac{\lambda}{4}$ . The voltage reflection coefficient  $\Gamma$  at the input is:

68.

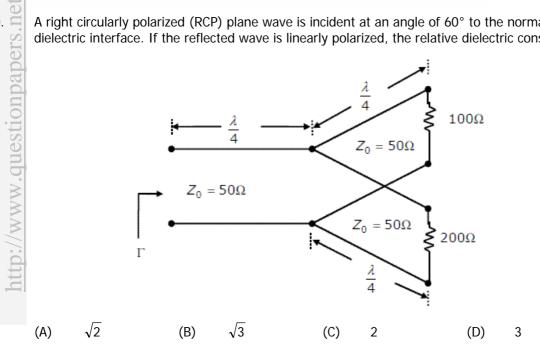


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dipole is kept horizontally at a height of  $\frac{\lambda_0}{2}$  above a perfectly conducting infinite ground plane. A  $\frac{\lambda}{2}$ 69. The radiation pattern in the plane of the dipole ( $\vec{E}$  plane) looks approximately as



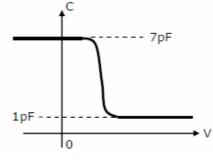
A right circularly polarized (RCP) plane wave is incident at an angle of 60° to the normal, on an airdielectric interface. If the reflected wave is linearly polarized, the relative dielectric constant  $\varepsilon_{r2}$  is:



#### **Common Data Questions**

#### Common Data for Questions 71, 72, 73:

The figure shows the high-frequency capacitance-voltage (C-V) characteristics of a Metal/SiO<sub>2</sub>/silicon (MOS) capacitor having an area of  $1 \times 10^{-4}$  cm<sup>2</sup>. Assume that the permittivities ( $\epsilon_0 \epsilon_r$ ) of silicon and SiO<sub>2</sub> are  $1 \times 10^{-12}$ F/cm and  $3.5 \times 10^{-13}$  F/cm respectively.



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71.	The g (A)	ate oxide thickn 50 nm	ess in the (B)	e MOS capacitor 143 nm	is: (C)	350 nm	(D)	1 µm
72.	The m (A)	naximum depleti 0.143 μm	on layer (B)	width in silicon is 0.857 μm	s (C)	1 μm	(D)	1.143 μm
73.	Consider the following statements about the C-V characteristics plot: S1: The MOS capacitor has an n-type substrate,							

S2: If positive charges are introduced in the oxide, the C-V plot will shift to the left.

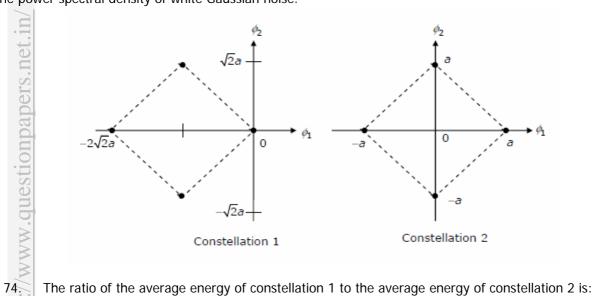
Then which of the following is true?

(A)	Both S1 and S2 are true	(B)	S1 is true and S2 is false

(C) S1 is false and S2 is true (D) Both S1 and S2 are false

#### Common Data for Questions 74, 75:

Two 4-ray signal constellations are shown. It is given that  $\phi_1$  and  $\phi_2$  constitute an orthonormal basis for the two constellations. Assume that the four symbols in both the constellations are equiprobable. Let  $\frac{N_0}{2}$  denote the power spectral density of white Gaussian noise.



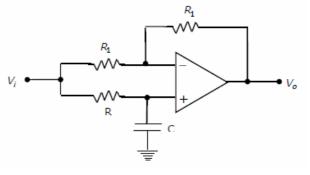
- The ratio of the average energy of constellation 1 to the average energy of constellation 2 is: (A)  $4a^2$  (B) 4 (C) 2 (D) 8
- 75. If these constellations are used for digital communications over an AWGN channel, then which of the following statements is true?
  - (A) Probability of symbol error for Constellation 1 is lower
  - (B) Probability of symbol error for Constellation 1 is higher
  - (C) Probability of symbol error is equal for both the constellations
  - (D) The value of  $N_0$  will determine which of the two constellations has a lower probability of symbol error.

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## Linked Answer Questions: Q.76 to Q.85 Carry Two Marks Each

## Statement for Linked Answer Questions 76 & 77

Consider the Op-Amp circuit shown in the figure.



76. The transfer function 
$$V_{b}(s)/V_{i}(s)$$
 is:  
(A)  $\frac{1-sRC}{1+sRC}$  (B)  $\frac{1+sRC}{1-sRC}$  (C)  $\frac{1}{1-sRC}$  (D)  $\frac{1}{1+sRC}$   
77. If  $V_{i} V_{i}(s)$  for (d) and  $V_{0} = V_{2}$  sin (at +  $\phi$ ), then the minimum and maximum values of  $\phi$  (in radians) are respectively  
(A)  $\frac{-\pi}{2}$  and  $\frac{\pi}{2}$  (B)  $0$  and  $\frac{\pi}{2}$  (C)  $-\pi$  and  $0$  (D)  $\frac{-\pi}{2}$  and  $0$   
Statement for Linked Answer Questions 78 & 79:  
An 8085 assembly language program is given below:  
Line 1: MVI B, 0EH  
3: XRI 60H  
3: KRI 60H  
3: KRI 60H  
3: KRI 60H  
3: KRI 60H  
4: ADD B  
5: ANI 9EH  
6: CP 19BH  
7: STA 3010H  
8: HIT  
78. The contents of the accumulator just after execution of ADD instruction in line 4 will be  
(A) C3H (B) EAH (C) DCH (D) 69H  
79. CY = 0, Z = 0 (B) CY = 0, Z = 1 (C) CY = 1, Z = 0 (D) CY = 1, Z = 1  
80. The eigenvalue and eigenvector pairs  $(\lambda_{0}, v)$  for the system are  
(A)  $\left(-1, \left[\frac{1}{-1}\right]\right)$  and  $\left(-2, \left[\frac{1}{-2}\right]\right)$  (B)  $\left(-1, \left[\frac{1}{-1}\right]\right)$  and  $\left(2, \left[\frac{1}{-2}\right]\right)$   
71. The system matrix A is:  
(A)  $\left[0, \left(\frac{1}{-1}, \frac{1}{-1}\right]\right]$  (B)  $\left[\frac{1}{-1}, \frac{1}{-2}\right]$  (C)  $\left[2, \frac{1}{-1}, \frac{1}{-1}\right]$  (D)  $\left[0, \frac{1}{-2}, \frac{1}{-3}\right]$   
51. Statement for Linked Answer Questions 82 & 83:  
An input to a 6-level quantizer has the probability density function f (x) as shown in the figure.  
Defision boundaries of the quantizer has the probability density tunction f (x) as shown in the figure.  
42. The values of a and b are:  
(A)  $a = \frac{1}{6}$  and  $b = \frac{1}{12}$  (B)  $a = \frac{1}{5}$  and  $b = \frac{3}{40}$   
(C)  $a = \frac{1}{4}$  and  $b = \frac{1}{16}$  (D)  $a = \frac{1}{3}$  and  $b = \frac{1}{24}$ 

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83. Assuming that the reconstruction levels of the quantizer are the mid-points of the decision boundaries, the ratio of signal power to quantization noise power is:

(A) 
$$\frac{152}{9}$$
 (B)  $\frac{64}{3}$  (C)  $\frac{76}{3}$  (D) 28

### Statement for Linked Answer Questions 84 & 85:

In the Digital-to-Analog converter circuit shown in the figure below,  $V_R = 10 \text{ V}$  and  $R = 10 \text{k}\Omega$ .

